

# **ELECTROTECNIA TEÓRICA**

Transparências das aulas teóricas

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# ELECTROTECNIA TEÓRICA

- Ondas electromagnéticas
- Linhas de transmissão

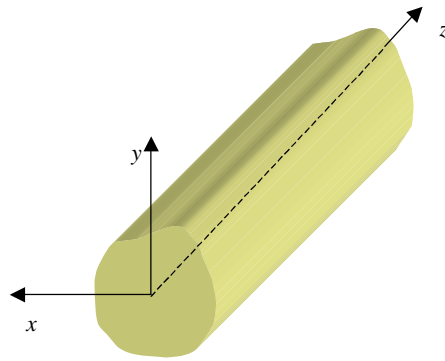
- Guias de onda cilíndricos
  - Guias metálicos
    - Placas paralelas
    - Rectangulares
    - Circulares
  - Guias dieléctricos
    - Planares
    - Fibras Ópticas

# GUIAS DE ONDA CILÍNDRICOS

## Guias de onda cilíndricos

- Formalismo teórico
  - Ondas guiadas
  - Método para obtenção de  $\vec{E}$  e  $\vec{H}$
  - Tipo de ondas
  - Frequência de corte
  - Impedância de onda
  - Potência média propagada
  - Energia média armazenada
  - Velocidade de transporte de energia
  - Condições fronteira

## Guias de onda cilíndricos



- secção transversal não varia com distância longitudinal
- guias preenchidos com material ( $\epsilon$ ,  $\mu$ ) sem perdas
- podem estar limitados por condutor perfeito ( $\sigma = \infty$ )
- comprimento infinito  $\Rightarrow$  propagação segundo +z



$$\vec{E}(x, y, z) = \vec{E}^0(x, y)e^{-\gamma z}$$
$$\vec{H}(x, y, z) = \vec{H}^0(x, y)e^{-\gamma z}$$

$$\nabla^2 \vec{E} + \omega^2 \mu \epsilon \vec{E} = 0$$
$$\nabla^2 \vec{H} + \omega^2 \mu \epsilon \vec{H} = 0$$

$$\nabla_{xy}^2 \vec{E}^0 + h^2 \vec{E}^0 = 0$$
$$\nabla_{xy}^2 \vec{H}^0 + h^2 \vec{H}^0 = 0$$

$$\nabla_{xy}^2 \vec{E}^0 + h^2 \vec{E}^0 = 0$$

$$\nabla_{xy}^2 \vec{H}^0 + h^2 \vec{H}^0 = 0$$



$$\frac{\partial^2 E_x^0}{\partial x^2} + \frac{\partial^2 E_x^0}{\partial y^2} + h^2 E_x^0 = 0$$

$$\frac{\partial^2 E_y^0}{\partial x^2} + \frac{\partial^2 E_y^0}{\partial y^2} + h^2 E_y^0 = 0$$

$$\frac{\partial^2 E_z^0}{\partial x^2} + \frac{\partial^2 E_z^0}{\partial y^2} + h^2 E_z^0 = 0$$

$$\frac{\partial^2 H_x^0}{\partial x^2} + \frac{\partial^2 H_x^0}{\partial y^2} + h^2 H_x^0 = 0$$

$$\frac{\partial^2 H_y^0}{\partial x^2} + \frac{\partial^2 H_y^0}{\partial y^2} + h^2 H_y^0 = 0$$

$$\frac{\partial^2 H_z^0}{\partial x^2} + \frac{\partial^2 H_z^0}{\partial y^2} + h^2 H_z^0 = 0$$

$$\begin{aligned}\nabla \times \vec{E} &= -j\omega \mathbf{m} \vec{H} \\ \nabla \times \vec{H} &= j\omega \mathbf{e} \vec{E}\end{aligned}$$



$$\begin{aligned}\frac{\partial E_z^0}{\partial y} + \mathbf{g}E_y^0 &= -j\omega \mathbf{m}H_x^0 & \frac{\partial H_z^0}{\partial y} + \mathbf{g}H_y^0 &= j\omega \mathbf{e}E_x^0 \\ -\frac{\partial E_z^0}{\partial x} - \mathbf{g}E_x^0 &= -j\omega \mathbf{m}H_y^0 & -\frac{\partial H_z^0}{\partial x} - \mathbf{g}H_x^0 &= j\omega \mathbf{e}E_y^0 \\ \frac{\partial E_y^0}{\partial x} - \frac{\partial E_x^0}{\partial y} &= -j\omega \mathbf{m}H_z^0 & \frac{\partial H_y^0}{\partial x} - \frac{\partial H_x^0}{\partial y} &= j\omega \mathbf{e}E_z^0\end{aligned} \quad \mathbf{e}$$



$$\begin{aligned}H_x^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial x} - j\omega \mathbf{e} \frac{\partial E_z^0}{\partial y} \right) \\ H_y^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial y} + j\omega \mathbf{e} \frac{\partial E_z^0}{\partial x} \right) \\ E_x^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial x} + j\omega \mathbf{m} \frac{\partial H_z^0}{\partial y} \right) \\ E_y^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial y} - j\omega \mathbf{m} \frac{\partial H_z^0}{\partial x} \right)\end{aligned}$$

## DETERMINAÇÃO DE $\vec{E}$ E $\vec{H}$

1. Resolver

$$\nabla_{xy}^2 E_z^0 + h^2 E_z^0 = 0$$

$$\nabla_{xy}^2 H_z^0 + h^2 H_z^0 = 0$$

$$h^2 = \mathbf{g}^2 + \mathbf{w}^2 \mathbf{m} \mathbf{e}$$

2. Calcular

$$H_x^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial x} - j\mathbf{w} \mathbf{e} \frac{\partial E_z^0}{\partial y} \right)$$

$$H_y^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial y} + j\mathbf{w} \mathbf{e} \frac{\partial E_z^0}{\partial x} \right)$$

$$E_x^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial x} + j\mathbf{w} \mathbf{m} \frac{\partial H_z^0}{\partial y} \right)$$

$$E_y^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial y} - j\mathbf{w} \mathbf{m} \frac{\partial H_z^0}{\partial x} \right)$$

3. Obter

$$\vec{E}(x, y, z) = \vec{E}^0(x, y) e^{-\mathbf{g}z}$$

$$\vec{H}(x, y, z) = \vec{H}^0(x, y) e^{-\mathbf{g}z}$$



$$h^2 = g^2 + w^2 me$$



$$g = \sqrt{h^2 - w^2 me} = w\sqrt{me} \sqrt{\frac{h^2}{w^2 me} - 1}$$

FREQUÊNCIA DE CORTE

$$f_c = \frac{h}{2p\sqrt{me}}$$



$$g = w\sqrt{me} \sqrt{\left(\frac{f_c}{f}\right)^2 - 1}$$

$$\mathbf{g} = w\sqrt{\mathbf{me}} \sqrt{\left(\frac{f_c}{f}\right)^2 - 1}$$

•  $f < f_c \Rightarrow \mathbf{g} = \mathbf{a} \Rightarrow$

$$\begin{aligned}\vec{E}(x, y, z) &= \vec{E}^0(x, y)e^{-\mathbf{a}z} \\ \vec{H}(x, y, z) &= \vec{H}^0(x, y)e^{-\mathbf{a}z}\end{aligned}$$



**modo evanescente**

•  $f > f_c \Rightarrow \mathbf{g} = \mathbf{jb} \Rightarrow$

$$\begin{aligned}\vec{E}(x, y, z) &= \vec{E}^0(x, y)e^{-\mathbf{jb}z} \\ \vec{H}(x, y, z) &= \vec{H}^0(x, y)e^{-\mathbf{jb}z}\end{aligned}$$



**modo em propagação**

## Modos em propagação

$$\mathbf{g} = j\mathbf{b}$$

$$\mathbf{b} = \mathbf{b}_m \sqrt{1 - \left(\frac{f_c}{f}\right)^2}, \quad \mathbf{b}_m = \omega \sqrt{\mathbf{m}\mathbf{e}}$$

$$\mathbf{l} = \frac{\mathbf{l}_m}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}}, \quad \mathbf{l}_m = \frac{2\mathbf{p}}{\mathbf{b}_m}$$

$$v_f = \frac{v_m}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}}, \quad v_m = \frac{1}{\sqrt{\mathbf{m}\mathbf{e}}}$$

$$v_g = v_m \sqrt{1 - \left(\frac{f_c}{f}\right)^2}$$

## Impedância de onda

- ondas planas propagando-se segundo +z num meio ilimitado

$$\begin{aligned} \vec{H} &= \frac{1}{h} (\hat{z} \times \vec{E}) \\ \vec{E} &= -h (\hat{z} \times \vec{H}) \end{aligned} \quad h = \sqrt{m/e}$$

- ondas guiadas

$$\begin{aligned} \text{ondas TEM ou TM: } \vec{H} &= \frac{1}{Z} (\hat{z} \times \vec{E}) \\ \text{ondas TEM ou TE: } \vec{E} &= -Z (\hat{z} \times \vec{H}) \end{aligned}$$

ou

$$\begin{aligned} H_x \hat{x} + H_y \hat{y} + H_z \hat{z} &= \frac{1}{Z} (-E_y \hat{x} + E_x \hat{y}) & \longrightarrow & H_z = 0 \\ E_x \hat{x} + E_y \hat{y} + E_z \hat{z} &= -Z (-H_y \hat{x} + H_x \hat{y}) & \longrightarrow & E_z = 0 \end{aligned}$$

⇓

$$Z = \frac{E_x}{H_y} = -\frac{E_y}{H_x}$$

**impedância  
de onda**

## Potência média propagada

$$P_{med} = \int_A \vec{S}_{med} \cdot d\vec{A}$$

$$d\vec{A} = dA \hat{z}$$

$$\vec{S}_{med} = \frac{1}{2} \text{Re}\{\vec{E} \times \vec{H}^*\}$$

$$P_{med} = \frac{1}{2} \int_A \text{Re}\{E_x H_y^* - E_y H_x^*\} dA$$

$$E_x = ZH_y$$

$$E_y = -ZH_x$$

$$P_{med} = \frac{1}{2} \int_A \text{Re}\left\{\frac{1}{Z}\right\} (|E_x|^2 + |E_y|^2) dA = \frac{1}{2} \int_A \text{Re}\{Z\} (|H_x|^2 + |H_y|^2) dA$$

## Energia média armazenada por unidade de comprimento

$$W'_{med} = \int_A (w_{e,med} + w_{m,med}) dA$$

$$w_{e,med} = \frac{\mathbf{e}}{4} \vec{E} \cdot \vec{E}^* = \frac{\mathbf{e}}{4} (|E_x|^2 + |E_y|^2 + |E_z|^2)$$

$$w_{m,med} = \frac{\mathbf{m}}{4} \vec{H} \cdot \vec{H}^* = \frac{\mathbf{m}}{4} (|H_x|^2 + |H_y|^2 + |H_z|^2).$$

## Velocidade de transporte de energia

$$v_{en} = \frac{P_{med}}{W'_{med}}$$

## Ondas TEM

$$\bullet \quad E_z = H_z = 0$$

$$\begin{aligned} E_x^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial x} + j\omega\mathbf{m} \frac{\partial H_z^0}{\partial y} \right) & H_x^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial x} - j\omega\mathbf{e} \frac{\partial E_z^0}{\partial y} \right) \\ E_y^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial y} - j\omega\mathbf{m} \frac{\partial H_z^0}{\partial x} \right) & H_y^0 &= -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial y} + j\omega\mathbf{e} \frac{\partial E_z^0}{\partial x} \right) \end{aligned}$$



$$h^2 = 0 \quad \Longrightarrow$$

$$f_c = \frac{h}{2p\sqrt{me}} = 0$$



$$\begin{aligned} \mathbf{g} &= j\mathbf{b} \\ \mathbf{b} &= \mathbf{b}_m \\ \mathbf{l} &= \mathbf{l}_m \\ v_f &= v_g = v_m \end{aligned}$$

$$\begin{aligned} \mathbf{g}E_y^0 &= -j\omega\mathbf{m}H_x^0 & \mathbf{g}H_y^0 &= j\omega\mathbf{e}E_x^0 \\ -\mathbf{g}E_x^0 &= -j\omega\mathbf{m}H_y^0 & -\mathbf{g}H_x^0 &= j\omega\mathbf{e}E_y^0 \\ \frac{\partial E_y^0}{\partial x} - \frac{\partial E_x^0}{\partial y} &= 0 & \frac{\partial H_y^0}{\partial x} - \frac{\partial H_x^0}{\partial y} &= 0 \end{aligned}$$



$$Z_{TEM} = \frac{j\omega\mathbf{m}}{\mathbf{g}} = \frac{\mathbf{g}}{j\omega\mathbf{e}} = \sqrt{\frac{\mathbf{m}}{\mathbf{e}}} = \mathbf{h}$$

## Ondas TM

$$\bullet E_z \neq 0 \text{ e } H_z = 0$$

$$\nabla_{xy}^2 E_z^0 + h^2 E_z^0 = 0$$

$$H_x^0 = \frac{j\omega \epsilon}{h^2} \frac{\partial E_z^0}{\partial y}$$

$$H_y^0 = -\frac{j\omega \epsilon}{h^2} \frac{\partial E_z^0}{\partial x}$$

$$E_x^0 = -\frac{g}{h^2} \frac{\partial E_z^0}{\partial x}$$

$$E_y^0 = -\frac{g}{h^2} \frac{\partial E_z^0}{\partial y}$$

$$Z_{TM} = \frac{g}{j\omega \epsilon} = \frac{\omega \sqrt{\mu \epsilon} \sqrt{\left(\frac{f_c}{f}\right)^2 - 1}}{j\omega \epsilon} = -j\mathbf{h} \sqrt{\left(\frac{f_c}{f}\right)^2 - 1}$$

- modos evanescentes:  $Z_{TM}$  imaginária  $\Rightarrow$  não há propagação de energia
- modos em propagação:  $Z_{TM} = \mathbf{h} \sqrt{1 - (f_c/f)^2}$  real e inferior a  $\mathbf{h}$



## Ondas TE

- $E_z = 0$  e  $H_z \neq 0$

$$\nabla_{xy}^2 H_z^0 + h^2 H_z^0 = 0$$

$$H_x^0 = -\frac{\mathbf{g}}{h^2} \frac{\partial H_z^0}{\partial x}$$

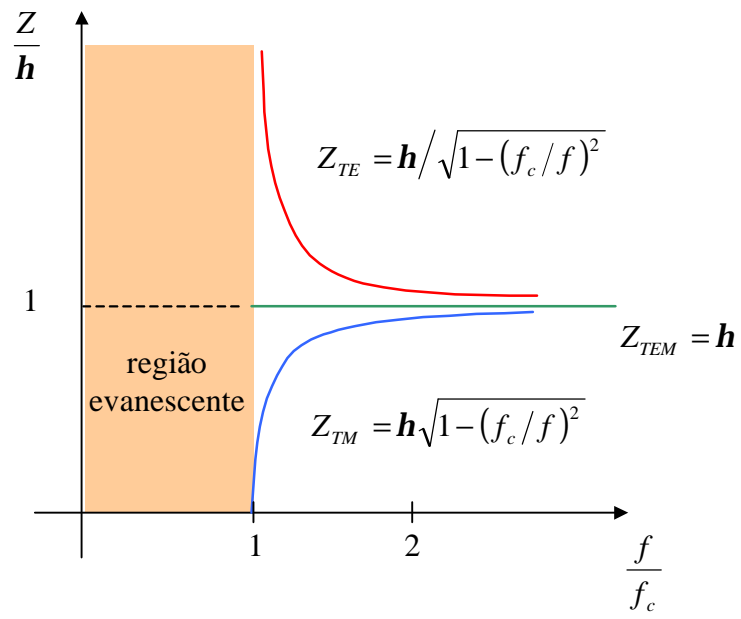
$$H_y^0 = -\frac{\mathbf{g}}{h^2} \frac{\partial H_z^0}{\partial y}$$

$$E_x^0 = -\frac{j\omega\mathbf{m}}{h^2} \frac{\partial H_z^0}{\partial y}$$

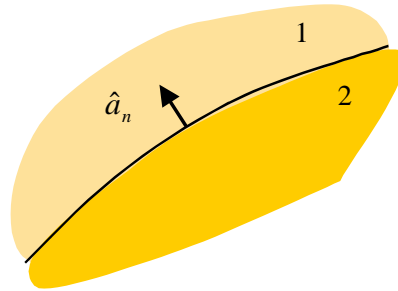
$$E_y^0 = \frac{j\omega\mathbf{m}}{h^2} \frac{\partial H_z^0}{\partial x}$$

$$Z_{TE} = \frac{j\omega\mathbf{m}}{\mathbf{g}} = \frac{j\mathbf{h}}{\sqrt{\left(\frac{f_c}{f}\right)^2 - 1}}$$

- modos evanescentes:  $Z_{TE}$  imaginária  $\implies$  não há propagação de energia
- modos em propagação:  $Z_{TE} = \mathbf{h} / \sqrt{1 - (f_c/f)^2}$  real e maior do que  $\mathbf{h}$



## Condições fronteira



$$\begin{array}{ll} \hat{a}_n \times (\vec{E}_1 - \vec{E}_2) = 0 & \hat{a}_n \cdot (\vec{B}_1 - \vec{B}_2) = 0 \\ \hat{a}_n \cdot (\vec{D}_1 - \vec{D}_2) = \mathbf{r}_s & \hat{a}_n \times (\vec{H}_1 - \vec{H}_2) = \vec{J}_s \end{array}$$



$$\begin{array}{l} E_{\text{tan}} \text{ contínuo} \\ B_{\text{norm}} \text{ contínuo} \\ D_{\text{norm}} \text{ contínuo se } \mathbf{r}_s = 0 \\ H_{\text{tan}} \text{ contínuo se } \vec{J}_s = 0 \end{array}$$

NOTAS:

- $\vec{D} = \mathbf{e} \vec{E} \quad \vec{B} = \mathbf{m} \vec{H}$
- $\vec{E}_{\text{cond}} = \vec{D}_{\text{cond}} = \vec{B}_{\text{cond}} = \vec{H}_{\text{cond}} = 0$

## GUIAS DE PLACAS PARALELAS

## Guias metálicos

- Guias limitados por condutores perfeitos

### Condições fronteira

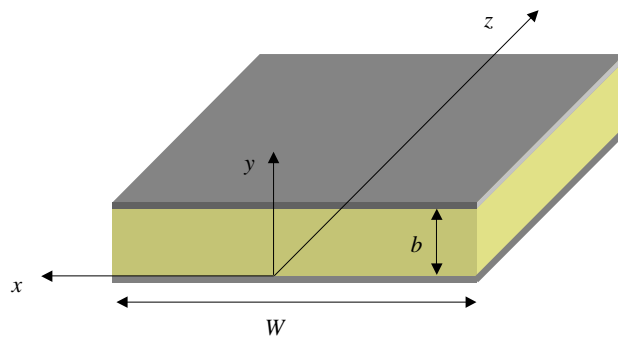
$E_{\text{tan}}$  e  $B_{\text{norm}}$  contínuos

$$E_{\text{condutor}} = B_{\text{condutor}} = 0$$



$$E_{\text{tan}} = H_{\text{normal}} = 0 \quad \text{junto aos condutores}$$

## Guias de placas paralelas



- $W \gg b \implies \frac{\partial}{\partial x} = 0$

## Condições fronteira

$$E_{\text{tan}} = H_{\text{normal}} = 0 \quad \text{em } y = 0 \text{ e } y = b$$



$$E_x^0 = E_z^0 = H_y^0 = 0 \quad \text{em } y = 0 \text{ e } y = b$$

## Ondas TEM

$$E_z^0 = H_z^0 = 0$$

$$\frac{\partial E_y^0}{\partial x} - \frac{\partial E_x^0}{\partial y} = 0 \quad \frac{\partial H_y^0}{\partial x} - \frac{\partial H_x^0}{\partial y} = 0$$

$$\partial/\partial x = 0$$

$$\frac{dE_x^0}{dy} = \frac{dH_x^0}{dy} = 0$$



$E_x^0$  e  $H_x^0$  são constantes

$$\begin{aligned} E_x^0(0) &= E_x^0(b) = 0 \\ E_x^0 &= Z_{TEM} H_y^0 \\ E_y^0 &= -Z_{TEM} H_x^0 \\ Z_{TEM} &= \mathbf{h} \end{aligned}$$

$$\begin{aligned} E_x^0 &= 0 \\ H_y^0 &= 0 \\ E_y^0 &= \text{constante} \end{aligned}$$

$$\begin{aligned} \vec{E}^0 &= E_0 \hat{y} \\ \vec{H}^0 &= -\frac{E_0}{\mathbf{h}} \hat{x} \end{aligned}$$

## Ondas TM

$$H_z^0 = 0 \quad \frac{d^2 E_z^0}{dy^2} + h^2 E_z^0 = 0$$

$$h \neq 0 \quad E_z^0(y) = A \sin(hy) + B \cos(hy)$$

$$E_z^0(0) = 0$$
$$E_z^0(b) = 0$$

$$B = 0$$
$$A \sin(hb) = 0 \Leftrightarrow hb = n\pi, \quad n = 1, 2, \dots$$

$$h = \frac{n\pi}{b}, \quad n = 1, 2, \dots$$
$$E_z^0 = A_n \sin\left(\frac{n\pi y}{b}\right)$$



$$h = \frac{n\mathbf{p}}{b}, \quad n=1, 2, \dots$$

$$E_z^0 = A_n \sin\left(\frac{n\mathbf{p} y}{b}\right)$$

$$H_x^0 = \frac{j\mathbf{w}e}{h^2} \frac{\partial E_z^0}{\partial y}$$

$$E_x^0 = -\frac{\mathbf{g}}{h^2} \frac{\partial E_z^0}{\partial x}$$

$$H_y^0 = -\frac{j\mathbf{w}e}{h^2} \frac{\partial E_z^0}{\partial x}$$

$$E_y^0 = -\frac{\mathbf{g}}{h^2} \frac{\partial E_z^0}{\partial y}$$

**MODO TM<sub>n</sub>**

$$E_z^0 = A_n \sin\left(\frac{n\mathbf{p} y}{b}\right)$$

$$H_x^0 = \frac{j\mathbf{w}eb}{n\mathbf{p}} A_n \cos\left(\frac{n\mathbf{p} y}{b}\right)$$

$$E_y^0 = -\frac{\mathbf{g}b}{n\mathbf{p}} A_n \cos\left(\frac{n\mathbf{p} y}{b}\right)$$

$$H_y^0 = E_x^0 = 0$$

$$\mathbf{g} = j\sqrt{\mathbf{w}^2 \mathbf{m}e - \left(\frac{n\mathbf{p}}{b}\right)^2}$$

## Ondas TE

$$E_z^0 = 0$$

$$\frac{d^2 H_z^0}{dy^2} + h^2 H_z^0 = 0$$



$$h \neq 0$$

$$H_z^0(y) = A \sin(hy) + B \cos(hy)$$

$$\begin{aligned} H_x^0 &= -\frac{\mathbf{g}}{h^2} \frac{\partial H_z^0}{\partial x} \\ E_x^0 &= -\frac{j\omega\mathbf{m}}{h^2} \frac{\partial H_z^0}{\partial y} \\ H_y^0 &= -\frac{\mathbf{g}}{h^2} \frac{\partial H_z^0}{\partial y} \\ E_y^0 &= \frac{j\omega\mathbf{m}}{h^2} \frac{\partial H_z^0}{\partial x} \end{aligned}$$



$$\begin{aligned} H_y^0 &= -\frac{\mathbf{g}}{h} [A \cos(hy) - B \sin(hy)] \\ E_x^0 &= -\frac{j\omega\mathbf{m}}{h} [A \cos(hy) - B \sin(hy)] \\ E_y^0 &= H_x^0 = 0 \end{aligned}$$

$$H_y^0 = -\frac{\mathbf{g}}{h} [A \cos(hy) - B \sin(hy)]$$

$$E_x^0 = -\frac{j\mathbf{w}\mathbf{m}}{h} [A \cos(hy) - B \sin(hy)]$$

$$E_y^0 = H_x^0 = 0$$

$$E_x^0(0) = E_x^0(b) = 0$$

$$H_y^0(0) = H_y^0(b) = 0$$

$$A = 0$$

$$\sin(hb) = 0 \Leftrightarrow h = \frac{n\mathbf{p}}{b}, \quad n = 1, 2, 3, \dots$$

### MODO TE<sub>n</sub>

$$H_z^0 = B_n \cos\left(\frac{n\mathbf{p} y}{b}\right)$$

$$H_y^0 = \frac{\mathbf{g} b}{n\mathbf{p}} B_n \sin\left(\frac{n\mathbf{p} y}{b}\right)$$

$$E_x^0 = j \frac{\mathbf{w}\mathbf{m} b}{n\mathbf{p}} B_n \sin\left(\frac{n\mathbf{p} y}{b}\right)$$

$$H_x^0 = E_y^0 = 0$$

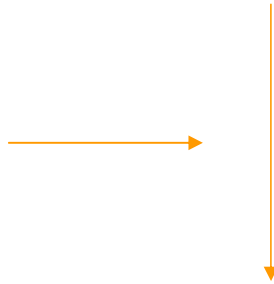
$$\mathbf{g} = j \sqrt{\mathbf{w}^2 \mathbf{m} \epsilon - \left(\frac{n\mathbf{p}}{b}\right)^2}$$

## Frequência de corte

$$f_c = \frac{h}{2p\sqrt{\epsilon}} \mathbf{m}$$

$$h_{TEM} = 0$$

$$h_{TM,TE} = \frac{n\mathbf{p}}{b}$$



$$(f_c)_{TEM} = 0$$

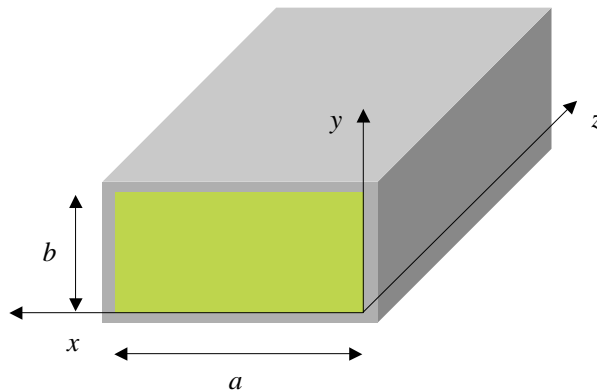
$$(f_c)_{TM,TE} = \frac{n}{2b\sqrt{\epsilon}} \mathbf{m}$$



**Modo dominante num guia de placas paralelas é o modo TEM**

# GUIAS RECTANGULARES

## Guias rectangulares



### Condições fronteira

$E_{\text{tan}}$  e  $B_{\text{norm}}$  contínuos

$$E_{\text{condutor}} = B_{\text{condutor}} = 0$$

$$\implies E_{\text{tan}} = H_{\text{normal}} = 0 \text{ junto ao condutor}$$



$$E_x^0 = E_z^0 = H_y^0 = 0 \text{ em } y = 0 \text{ ou } y = b$$

$$E_y^0 = E_z^0 = H_x^0 = 0 \text{ em } x = 0 \text{ ou } x = a$$



$$\frac{d^2 X(x)}{dx^2} + k_x^2 X(x) = 0 \quad \longrightarrow \quad X(x) = A \sin(k_x x) + B \cos(k_x x)$$

$$\frac{d^2 Y(y)}{dy^2} + k_y^2 Y(y) = 0 \quad \longrightarrow \quad Y(y) = C \sin(k_y y) + D \cos(k_y y)$$



$$y(x, y) = [A \sin(k_x x) + B \cos(k_x x)][C \sin(k_y y) + D \cos(k_y y)]$$


$$h^2 = k_x^2 + k_y^2$$



## Ondas TM


$$H_z^0 = 0$$

$$\nabla_{xy}^2 E_z^0 + h^2 E_z^0 = 0$$


$$E_z^0 = \underbrace{[A \sin(k_x x) + B \cos(k_x x)]}_{X(x)} \underbrace{[C \sin(k_y y) + D \cos(k_y y)]}_{Y(y)}$$

$$E_z^0(0, y) = E_z^0(a, y) = 0 \quad \longrightarrow \quad B = 0 \quad \text{e} \quad A \sin(k_x a) = 0$$

$$E_z^0(x, 0) = E_z^0(x, b) = 0 \quad \longrightarrow \quad D = 0 \quad \text{e} \quad C \sin(k_y b) = 0$$


$$k_x = \frac{m\pi}{a}, \quad m \text{ inteiro} \quad \text{e} \quad X(x) = A \sin\left(\frac{m\pi x}{a}\right)$$
$$k_y = \frac{n\pi}{b}, \quad n \text{ inteiro} \quad \text{e} \quad Y(y) = C \sin\left(\frac{n\pi y}{b}\right)$$

### modo $\text{TM}_{mn}$

$$E_z^0 = E_{0,mn} \sin\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$$

$$H_x^0 = \frac{j\omega\epsilon}{h^2} \frac{n\pi}{b} E_{0,mn} \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right)$$

$$H_y^0 = -\frac{j\omega\epsilon}{h^2} \frac{m\pi}{a} E_{0,mn} \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$$

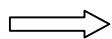
$$E_x^0 = -\frac{\gamma}{h^2} \frac{m\pi}{a} E_{0,mn} \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$$

$$E_y^0 = -\frac{\gamma}{h^2} \frac{n\pi}{b} E_{0,mn} \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right)$$

$$h^2 = \left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2$$

Importante:

- $h \neq 0 \Rightarrow m \neq 0$  ou  $n \neq 0$
- $n=0$  ou  $m=0 \Rightarrow \vec{E} = \vec{H} = 0$



$$n \geq 1 \text{ e } m \geq 1$$

## Ondas TE

$$E_z^0 = 0$$

$$\nabla_{xy}^2 H_z^0 + h^2 H_z^0 = 0$$

$$H_z^0 = \underbrace{[A \sin(k_x x) + B \cos(k_x x)]}_{X(x)} \underbrace{[C \sin(k_y y) + D \cos(k_y y)]}_{Y(y)}$$

$$E_x^0 = -j \frac{\omega \mu}{h^2} X(x) Y'(y)$$

$$E_y^0 = j \frac{\omega \mu}{h^2} X'(x) Y(y)$$

onde

$$X'(x) = k_x [A \cos(k_x x) - B \sin(k_x x)]$$

$$Y'(y) = k_y [C \cos(k_y y) - D \sin(k_y y)]$$

$$E_y^0(0, y) = E_y^0(a, y) = 0$$



$$A = 0 \quad \text{e} \quad k_x = \frac{m\pi}{a}, \quad m \text{ inteiro}$$

$$E_x^0(x, 0) = E_x^0(x, b) = 0$$



$$C = 0 \quad \text{e} \quad k_y = \frac{n\pi}{b}, \quad n \text{ inteiro}$$

### modo $TE_{mn}$

$$H_z^0 = H_{0,mn} \cos\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right)$$

$$H_x^0 = \frac{\gamma}{h^2} \frac{m\pi}{a} H_{0,mn} \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right)$$

$$H_y^0 = \frac{\gamma}{h^2} \frac{n\pi}{b} H_{0,mn} \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$$

$$E_x^0 = \frac{j\omega\mu}{h^2} \frac{n\pi}{b} H_{0,mn} \cos\left(\frac{m\pi x}{a}\right) \sin\left(\frac{n\pi y}{b}\right)$$

$$E_y^0 = -\frac{j\omega\mu}{h^2} \frac{m\pi}{a} H_{0,mn} \sin\left(\frac{m\pi x}{a}\right) \cos\left(\frac{n\pi y}{b}\right)$$

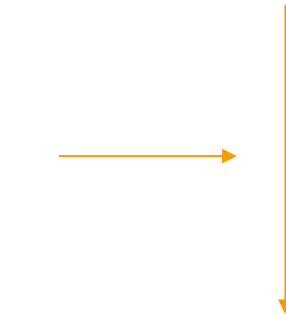
$$h^2 = \left(\frac{m\mathbf{p}}{a}\right)^2 + \left(\frac{n\mathbf{p}}{b}\right)^2$$

- $h \neq 0 \Rightarrow m \neq 0$  ou  $n \neq 0$

## Frequência de corte

$$h = \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

$$f_c = \frac{h}{2\pi\sqrt{\mu\epsilon}}$$



$$f_c = \frac{1}{2\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m}{a}\right)^2 + \left(\frac{n}{b}\right)^2}$$

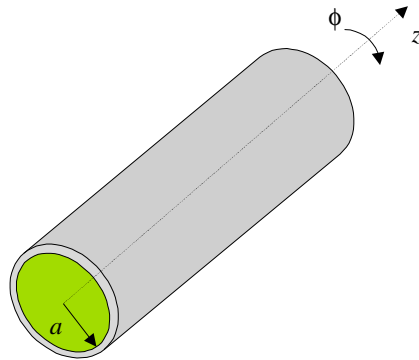
- modos TM:  $n \geq 1$  e  $m \geq 1 \implies$  modo TM com frequência de corte mais baixa é modo  $TM_{11}$
- modos TE:  $m \neq 0$  ou  $n \neq 0 \implies$  se  $a > b$ , modo TE com a frequência de corte mais baixa é o modo  $TE_{10}$

$$(f_c)_{TE_{10}} < (f_c)_{TM_{11}} \implies$$

**modo  $TE_{10}$  é modo dominante nos guias rectangulares**

# GUIAS CIRCULARES

## Guias circulares



coordenadas cilíndricas



$$\begin{aligned}\vec{E} &= (E_r^0 \hat{r} + E_\phi^0 \hat{\phi} + E_z^0 \hat{z}) e^{-g z} \\ \vec{H} &= (H_r^0 \hat{r} + H_\phi^0 \hat{\phi} + H_z^0 \hat{z}) e^{-g z}\end{aligned}$$

## Condições fronteira

$$E_{\text{tan}} = H_{\text{normal}} = 0 \text{ junto ao condutor}$$



$$E_\phi^0 = E_z^0 = H_r^0 = 0 \text{ em } r = a$$

## DETERMINAÇÃO DE $\vec{E}$ E $\vec{H}$

1. Resolver as equações

$$\nabla_{rf}^2 E_z^0 + h^2 E_z^0 = 0$$

$$\nabla_{rf}^2 H_z^0 + h^2 H_z^0 = 0$$

$$h^2 = \mathbf{g}^2 + \mathbf{w}^2 \mathbf{m} \mathbf{e}$$

$$\nabla_{rf}^2 = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2}{\partial \mathbf{f}^2}$$

2. Calcular

$$H_r^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial H_z^0}{\partial r} - \frac{j\mathbf{w}\mathbf{e}}{r} \frac{\partial E_z^0}{\partial \mathbf{f}} \right)$$

$$H_f^0 = -\frac{1}{h^2} \left( \frac{\mathbf{g}}{r} \frac{\partial H_z^0}{\partial \mathbf{f}} + j\mathbf{w}\mathbf{e} \frac{\partial E_z^0}{\partial r} \right)$$

$$E_r^0 = -\frac{1}{h^2} \left( \mathbf{g} \frac{\partial E_z^0}{\partial r} + \frac{j\mathbf{w}\mathbf{m}}{r} \frac{\partial H_z^0}{\partial \mathbf{f}} \right)$$

$$E_f^0 = -\frac{1}{h^2} \left( \frac{\mathbf{g}}{r} \frac{\partial E_z^0}{\partial \mathbf{f}} - j\mathbf{w}\mathbf{m} \frac{\partial H_z^0}{\partial r} \right)$$



## Equação de onda em coordenadas cilíndricas

$$\nabla_{r\mathbf{f}}^2 \mathbf{y} + h^2 \mathbf{y} = 0$$



$$\frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial \mathbf{y}}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 \mathbf{y}}{\partial \mathbf{f}^2} + h^2 \mathbf{y} = 0$$

$$\mathbf{y}(r, \mathbf{f}) = R(r)\Phi(\mathbf{f})$$

$$k_{\mathbf{f}}^2 \left\langle \underbrace{\frac{r^2}{R(r)} \frac{d^2 R(r)}{dr^2} + \frac{r}{R(r)} \frac{dR(r)}{dr} + h^2 r^2}_{\frac{\partial}{\partial \mathbf{f}} = 0} \right\rangle = \left\langle \underbrace{-\frac{1}{\Phi(\mathbf{f})} \frac{d^2 \Phi(\mathbf{f})}{d\mathbf{f}^2}}_{\frac{\partial}{\partial r} = 0} \right\rangle k_{\mathbf{f}}^2$$

$$\frac{d^2\Phi(\mathbf{f})}{d\mathbf{f}^2} + k_f^2\Phi(\mathbf{f}) = 0$$



$$\Phi(\mathbf{f}) = A\sin(k_f\mathbf{f}) + B\cos(k_f\mathbf{f})$$



$$\begin{aligned} \Phi(\mathbf{f} + 2\mathbf{p}) &= \Phi(\mathbf{f}) \\ \Downarrow \\ \sin(k_f\mathbf{f} + k_f2\mathbf{p}) &= \sin(k_f\mathbf{f}) \\ \cos(k_f\mathbf{f} + k_f2\mathbf{p}) &= \cos(k_f\mathbf{f}) \\ \Downarrow \\ k_f &= n, \quad n \text{ inteiro} \end{aligned}$$



$$\Phi(\mathbf{f}) = A\sin(n\mathbf{f}) + B\cos(n\mathbf{f})$$



$$\Phi(\mathbf{f}) = B\cos(n\mathbf{f})$$

$$\frac{r^2}{R(r)} \frac{d^2 R(r)}{dr^2} + \frac{r}{R(r)} \frac{dR(r)}{dr} + h^2 r^2 = n^2$$



$$r^2 \frac{d^2 R(r)}{dr^2} + r \frac{dR(r)}{dr} + (h^2 r^2 - n^2) R(r) = 0$$

**equação diferencial  
de Bessel**



$$R(r) = C J_n(hr) + D N_n(hr)$$

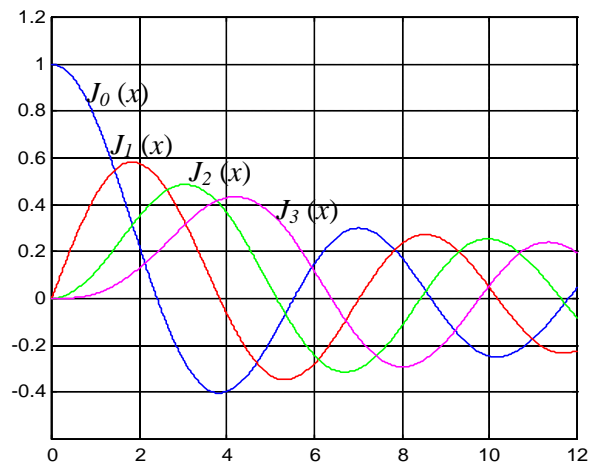
funções de  
Bessel de 1<sup>a</sup>  
espécie

funções de  
Bessel de 2<sup>a</sup>  
espécie

## Funções de Bessel de 1ª espécie

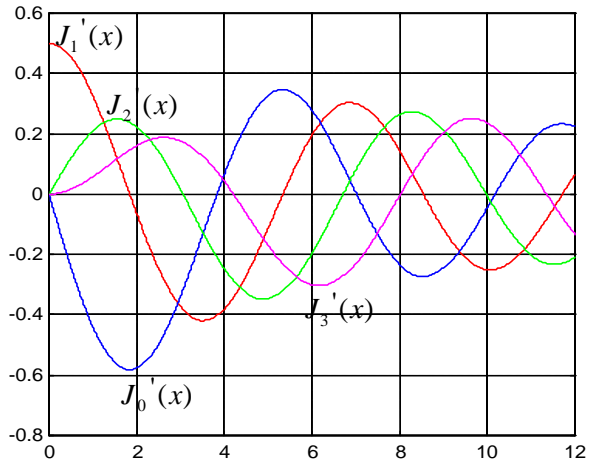
para  $n$  inteiro

$$J_n(x) = \sum_{m=0}^{\infty} \frac{(-1)^m x^{n+2m}}{m!(m+n)!2^{n+2m}}$$



zero	$J_0(x)$	$J_1(x)$	$J_2(x)$	$J_3(x)$
1	2.4048	3.8317	5.1336	6.3802
2	5.5201	7.0156	8.4172	9.7610
3	8.6537	10.1735	11.6198	13.0152
4	11.7915	13.3237	14.7960	16.2235
5	14.9309	16.4706	17.9598	19.4094

$$J'_n(x) = \frac{1}{2}[J_{n-1}(x) - J_{n+1}(x)]$$

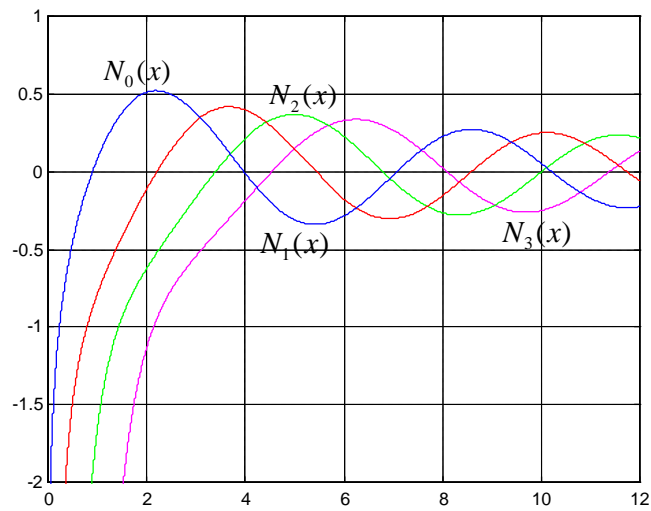


zero	$J'_0(x)$	$J'_1(x)$	$J'_2(x)$	$J'_3(x)$
1	3.8317	1.8412	3.0542	4.2012
2	7.0156	5.3314	6.7061	8.0152
3	10.1735	8.5363	9.9695	11.3459
4	13.3237	11.7060	13.1704	14.5858
5	16.4706	14.8636	16.3475	17.7887

## Funções de Bessel de 2ª espécie

para  $n$  inteiro

$$N_n(x) = \lim_{p \rightarrow n} \frac{J_p(x) \cos(pp) - J_{-p}(x)}{\sin(pp)}$$



$$R(r) = C J_n(hr) + D N_n(hr)$$

toma valores  
infinitos em  $r=0$



$$D=0$$

$$R(r) = C J_n(hr)$$



$$\nabla_{r\mathbf{f}}^2 \mathbf{y} + h^2 \mathbf{y} = 0$$



$$\mathbf{y}(r, \mathbf{f}) = R(r)\Phi(\mathbf{f}) = C_n J_n(hr) \cos(n\mathbf{f})$$

## Ondas TM

$$H_z^0 = 0$$

$$\nabla_{rf}^2 E_z^0 + h^2 E_z^0 = 0$$

$$E_z^0 = C_n J_n(hr) \cos(nf)$$

$$H_r^0 = \frac{j\omega \epsilon}{h^2 r} \frac{\partial E_z^0}{\partial f}$$

$$H_f^0 = -\frac{j\omega \epsilon}{h^2} \frac{\partial E_z^0}{\partial r}$$

$$E_r^0 = -\frac{g}{h^2} \frac{\partial E_z^0}{\partial r}$$

$$E_f^0 = -\frac{g}{h^2 r} \frac{\partial E_z^0}{\partial f}$$

$$H_r^0 = -\frac{j\omega \epsilon n}{h^2 r} C_n J_n(hr) \sin(nf)$$

$$H_f^0 = -\frac{j\omega \epsilon}{h} C_n J'_n(hr) \cos(nf)$$

$$E_r^0 = -\frac{g}{h} C_n J'_n(hr) \cos(nf)$$

$$E_f^0 = \frac{g n}{h^2 r} C_n J_n(hr) \sin(nf)$$

$$E_z^0(r = a, f) = 0 \quad \Rightarrow \quad J_n(ha) = 0$$



$$J_n(ha) = 0$$



$$\begin{aligned} n=0 &\rightarrow h = \frac{2.4048}{a}; h = \frac{5.5201}{a}; h = \frac{8.6537}{a}; \dots \\ n=1 &\rightarrow h = \frac{3.8317}{a}; h = \frac{7.0156}{a}; h = \frac{10.1735}{a}; \dots \\ &\vdots \end{aligned}$$

modo  $TM_{np}$   $\Rightarrow$

$$h = h_{TM_{np}} = \frac{p - \text{ésimo zero de } J_n}{a}$$

### Frequência de corte

$$(f_c)_{TM_{np}} = \frac{h_{TM_{np}}}{2p \sqrt{\mu \epsilon}} = \frac{p - \text{ésimo zero de } J_n}{2p a \sqrt{\mu \epsilon}}$$

menor zero de  $J_n$  é  
2.4048 ( $n=0, p=1$ )  $\Rightarrow$

modo TM dominante é  
o modo  $TM_{01}$

$$(f_c)_{TM_{01}} = \frac{2.4048}{2p a \sqrt{\mu \epsilon}}$$

## Ondas TE

$$E_z^0 = 0$$

$$\nabla_{rf}^2 H_z^0 + h^2 H_z^0 = 0$$

$$H_z^0 = C_n J_n(hr) \cos(nf)$$

$$\begin{aligned} H_r^0 &= -\frac{\mathbf{g}}{h^2} \frac{\partial H_z^0}{\partial r} \\ H_f^0 &= -\frac{\mathbf{g}}{h^2 r} \frac{\partial H_z^0}{\partial f} \\ E_r^0 &= -\frac{j\omega\mathbf{m}}{h^2 r} \frac{\partial H_z^0}{\partial f} \\ E_f^0 &= \frac{j\omega\mathbf{m}}{h^2} \frac{\partial H_z^0}{\partial r} \end{aligned}$$

$$\begin{aligned} H_r^0 &= -\frac{\mathbf{g}}{h} C_n J'_n(hr) \cos(nf) \\ H_f^0 &= \frac{\mathbf{g}n}{h^2 r} C_n J_n(hr) \sin(nf) \\ E_r^0 &= \frac{j\omega\mathbf{m}n}{h^2 r} C_n J_n(hr) \sin(nf) \\ E_f^0 &= \frac{j\omega\mathbf{m}}{h} C_n J'_n(hr) \cos(nf) \end{aligned}$$

$$E_f^0(r = a, f) = 0 \quad \Rightarrow \quad J'_n(ha) = 0$$

$$J'_n(ha) = 0$$



$$\begin{aligned} n=0 &\rightarrow h = \frac{3.8317}{a}; h = \frac{7.0156}{a}; \dots \\ n=1 &\rightarrow h = \frac{1.8412}{a}; h = \frac{5.3314}{a}; \dots \\ &\vdots \end{aligned}$$

modo  $TE_{np}$



$$h = h_{TE_{np}} = \frac{p - \text{ésimo zero de } J'_n}{a}$$

### Frequência de corte

$$(f_c)_{TE_{np}} = \frac{h_{TE_{np}}}{2p\sqrt{\mu\epsilon}} = \frac{p - \text{ésimo zero de } J'_n}{2p a \sqrt{\mu\epsilon}}$$

menor zero de  $J'_n$  é  
1.8412 ( $n=1, p=1$ )



modo TE dominante é  
o modo  $TE_{11}$

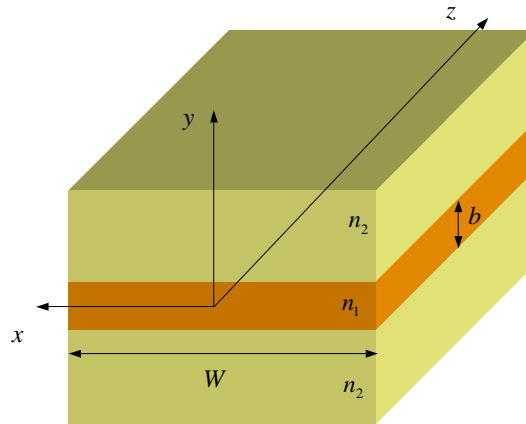
$$(f_c)_{TE_{11}} = \frac{1.8412}{2p a \sqrt{\mu\epsilon}}$$

$$(f_c)_{TE_{11}} < (f_c)_{TM_{01}} \implies$$

**modo dominante  
num guia circular é  
o modo  $TE_{11}$ !**

# GUIAS DIELECTRICOS PLANARES

## Guias dielétricos planares



$$\bullet \quad W \gg b \implies \frac{\partial}{\partial x} = 0$$

### Condições fronteira

$$\mathbf{r}_s = 0$$

$$\vec{\mathbf{J}}_s = 0$$



$E_{\text{tan}}$	contínua
$B_{\text{normal}}$	contínua
$D_{\text{normal}}$	contínua
$H_{\text{tan}}$	contínua

## Equação de onda em guias dielétricos planares

$$\frac{d^2 \mathbf{y}}{dy^2} + h^2 \mathbf{y} = 0 \quad h^2 = \begin{cases} h_1^2, & |y| \leq \frac{b}{2} \\ h_2^2, & |y| > \frac{b}{2} \end{cases}$$

- $h^2 > 0 \Rightarrow h \text{ real} \Rightarrow \mathbf{y} = A \sin(hy) + B \cos(hy)$
- $h^2 < 0 \Rightarrow h = j\mathbf{n} \Rightarrow \mathbf{y} = C e^{-\mathbf{n}y} + D e^{+\mathbf{n}y}$



$$\begin{aligned} h_1 & \text{ real} \\ h_2 & = j\mathbf{n} \end{aligned}$$

$$\begin{cases} h_1^2 = \mathbf{g}^2 + \left(\frac{\mathbf{w}}{c} n_1\right)^2 \\ \mathbf{n}^2 = -\mathbf{g}^2 - \left(\frac{\mathbf{w}}{c} n_2\right)^2 \end{cases} \Rightarrow \mathbf{n} = \sqrt{\left(\frac{\mathbf{w}}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2}$$

$$\bullet \quad \mathbf{g} = j\mathbf{b} \quad \longrightarrow \quad \mathbf{b} = \sqrt{\left(\frac{\mathbf{w}}{c} n_1\right)^2 - h_1^2} = \sqrt{\left(\frac{\mathbf{w}}{c} n_2\right)^2 + \mathbf{n}^2}$$



$$n_1 > n_2$$



$$\frac{\mathbf{w}}{c} n_1 > \mathbf{b} > \frac{\mathbf{w}}{c} n_2$$

$$\mathbf{y}(y) = \begin{cases} De^{ny}, & y < -\frac{b}{2} \\ A \sin(h_1 y) + B \cos(h_1 y), & |y| \leq \frac{b}{2} \\ Ce^{-ny}, & y > \frac{b}{2} \end{cases}$$

$\mathbf{y}(y)$  contínua  
em  $y = \pm b/2$

$$\mathbf{y}(y) = \begin{cases} \left[ -A \sin\left(\frac{h_1 b}{2}\right) + B \cos\left(\frac{h_1 b}{2}\right) \right] e^{n\left(y + \frac{b}{2}\right)}, & y < -\frac{b}{2} \\ A \sin(h_1 y) + B \cos(h_1 y), & |y| \leq \frac{b}{2} \\ \left[ A \sin\left(\frac{h_1 b}{2}\right) + B \cos\left(\frac{h_1 b}{2}\right) \right] e^{-n\left(y - \frac{b}{2}\right)}, & y > \frac{b}{2} \end{cases}$$

**modos pares**

$$\mathbf{y}_{par} = \begin{cases} B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y + \frac{b}{2}\right)}, & y < -\frac{b}{2} \\ B \cos(h_1 y), & |y| \leq \frac{b}{2} \\ B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y - \frac{b}{2}\right)}, & y > \frac{b}{2} \end{cases}$$

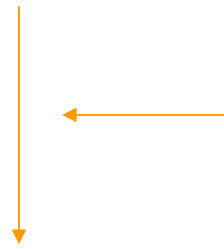
**modos ímpares**

$$\mathbf{y}_{impar} = \begin{cases} A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y + \frac{b}{2}\right)}, & y < -\frac{b}{2} \\ A \sin(h_1 y), & |y| \leq \frac{b}{2} \\ A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y - \frac{b}{2}\right)}, & y > \frac{b}{2} \end{cases}$$

## Ondas TM pares

$$H_z^0 = 0$$

$$E_z^0 = \mathbf{y}_{par}$$



$H_x^0 = \frac{j\omega\epsilon}{h^2} \frac{\partial E_z^0}{\partial y}$ $E_y^0 = -\frac{\mathbf{g}}{h^2} \frac{\partial E_z^0}{\partial y}$ $E_x^0 = H_y^0 = 0$
---

$$|y| \leq \frac{b}{2} :$$

$$\left. \begin{aligned} E_z^0 &= B \cos(h_1 y) \\ H_x^0 &= -\frac{j\omega\epsilon_1}{h_1} B \sin(h_1 y) \\ E_y^0 &= \frac{j\mathbf{b}}{h_1} B \sin(h_1 y) \end{aligned} \right\}$$

$$y > \frac{b}{2} :$$

$$\left. \begin{aligned} E_z^0 &= B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\ H_x^0 &= \frac{j\omega\epsilon_2}{n} B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\ E_y^0 &= -\frac{j\mathbf{b}}{n} B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \end{aligned} \right\}$$

$$y < -\frac{b}{2} :$$

$$\left. \begin{aligned} E_z^0 &= B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\ H_x^0 &= -\frac{j\omega\epsilon_2}{n} B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\ E_y^0 &= \frac{j\mathbf{b}}{n} B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \end{aligned} \right\}$$

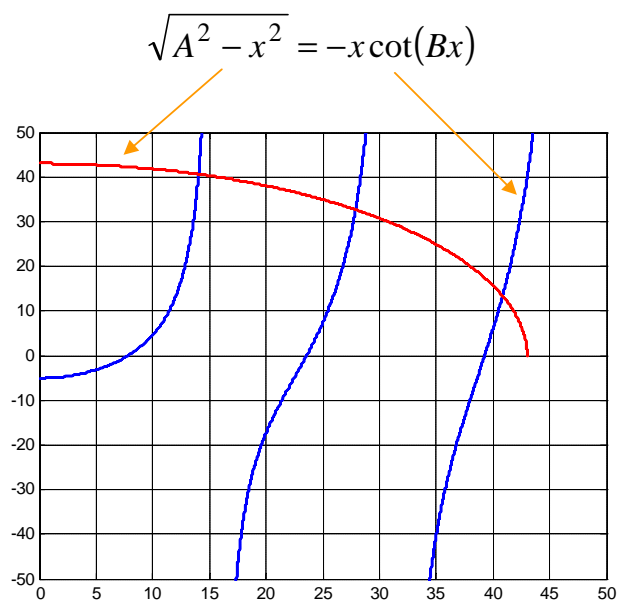


$\mathbf{n} = -h_1 \left(\frac{n_2}{n_1}\right)^2 \cot\left(\frac{h_1 b}{2}\right)$
---



$$\left(\frac{n_1}{n_2}\right)^2 \sqrt{\left(\frac{w}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2} = -h_1 \cot\left(\frac{h_1 b}{2}\right)$$

**EQUAÇÃO  
CARACTERÍSTICA**



valores característicos ( $h_1$ ) em guias dielétricos planares:

- **número finito** (correspondendo cada valor a um modo que se pode propagar no guia à frequência considerada);
- **dependem da frequência** de operação (valor  $A^2$  é proporcional à frequência de operação  $w$ ).

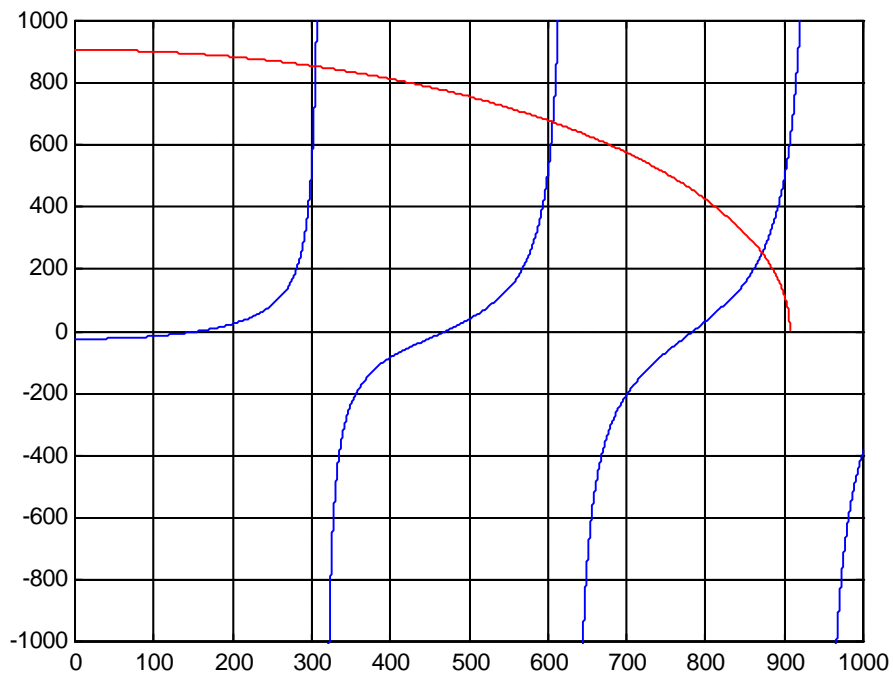
## EXEMPLO

$$f = 25 \text{ GHz}$$

$$\mathbf{e} = 4\mathbf{e}_0$$

$$\mathbf{m} = \mathbf{m}_0$$

$$b = 2 \text{ cm}$$

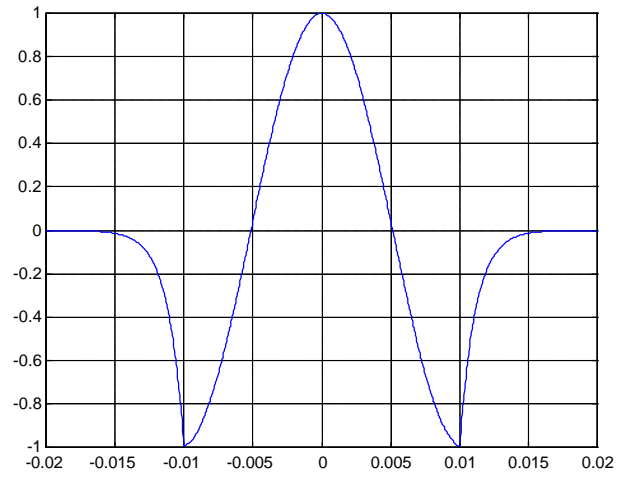


$$h_{1,1} = 305.25$$

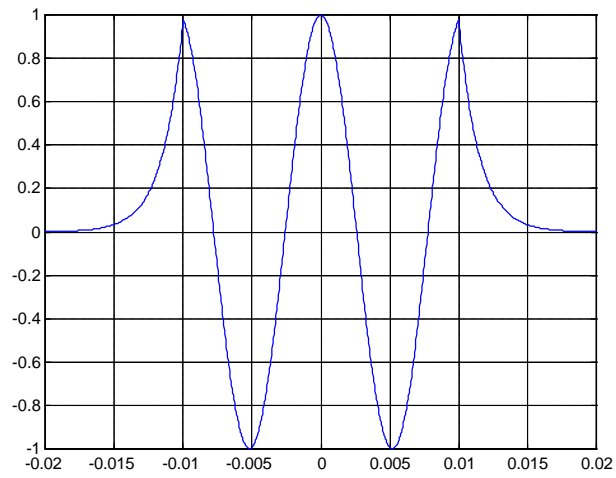
$$h_{1,2} = 606.22$$

$$h_{1,3} = 871.20$$

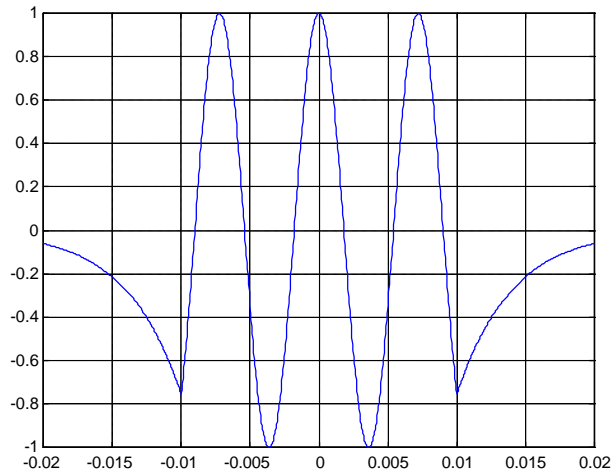
$$TM_{par,1}$$
$$h_{1,1} = 305.25$$
$$n_1 = 853.98$$



$$TM_{par,2}$$
$$h_{1,2} = 606.22$$
$$n_2 = 674.51$$



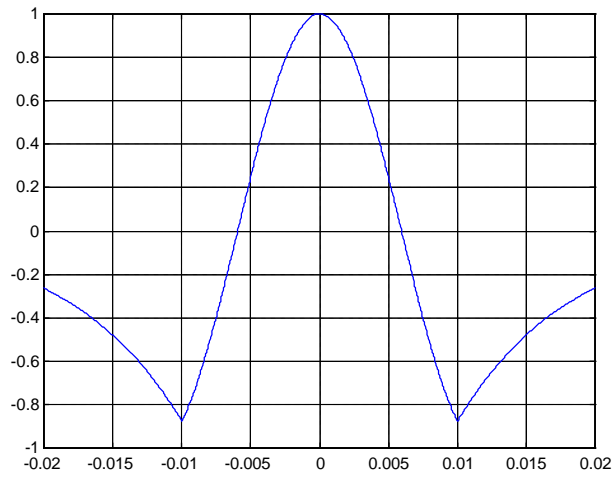
$$TM_{par,3}$$
$$h_{1,3} = 871.20$$
$$n_3 = 251.98$$



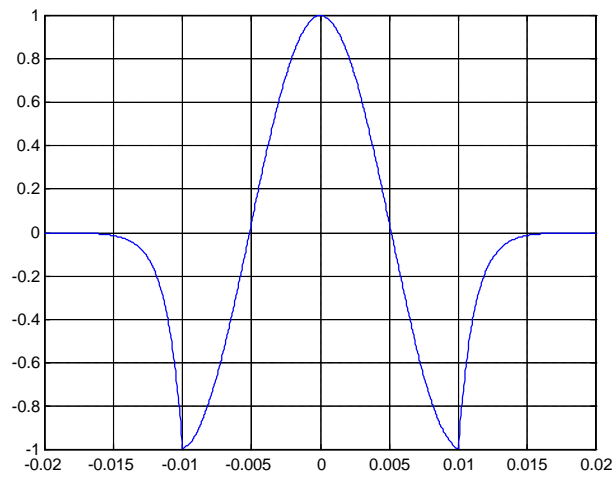
# $TM_{par,1}$

---

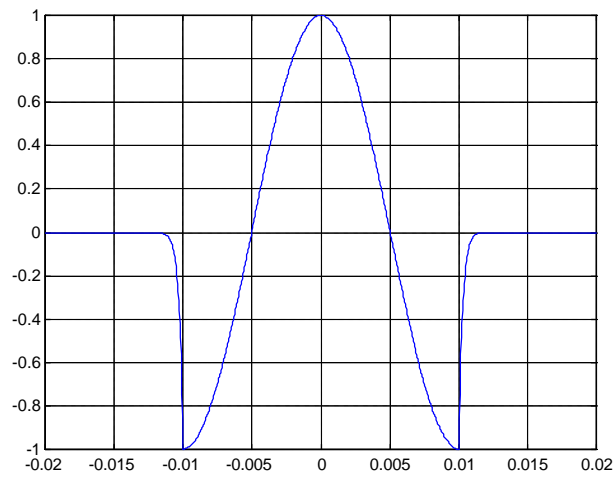
$f = 8 \text{ GHz}$   
 $h_{d,1} = 264.03$   
 $n = 120.45$



$f = 25 \text{ GHz}$   
 $h_{d,1} = 305.25$   
 $n = 853.98$



$f = 100 \text{ GHz}$   
 $h_{d,1} = 312$   
 $n = 3614.16$



## Frequência de corte

Condição de corte:

$$n = 0$$

- para o corte:

$$n^2 = -g^2 - \left(\frac{w}{c}n_2\right)^2 = 0 \Leftrightarrow g^2 = -\left(\frac{w}{c}n_2\right)^2$$

$$h_1^2 = g^2 + \left(\frac{w}{c}n_1\right)^2 = \left(\frac{w}{c}\right)^2(n_1^2 - n_2^2) \Leftrightarrow h_1 = \frac{w}{c}\sqrt{n_1^2 - n_2^2}$$

$$n = -h_1\left(\frac{n_2}{n_1}\right)^2 \cot\left(\frac{h_1 b}{2}\right) = 0 \Leftrightarrow -\cot\left(\frac{h_1 b}{2}\right) = 0$$



$$\frac{h_1 b}{2} = \frac{w b \sqrt{n_1^2 - n_2^2}}{2c} = \left(n - \frac{1}{2}\right) p, \quad n = 1, 2, \dots$$



$$(f_c)_{TM \text{ par}} = \frac{\left(n - \frac{1}{2}\right)c}{b\sqrt{n_1^2 - n_2^2}}, \quad n = 1, 2, \dots$$



frequência de corte  
aumenta com a diminuição  
da largura do guia!

## Modos TM ímpares

$$H_z^0 = 0$$

$$E_z^0 = \mathbf{y}_{\text{ímpar}}$$

$$\begin{array}{l}
 |y| \leq \frac{b}{2}: \\
 \\
 y > \frac{b}{2}: \\
 \\
 y < -\frac{b}{2}:
 \end{array}
 \left|
 \begin{array}{l}
 E_z^0 = A \sin(h_1 y) \\
 H_x^0 = \frac{j\mathbf{w}\mathbf{e}_1}{h_1} A \cos(h_1 y) \\
 E_y^0 = -\frac{j\mathbf{b}}{h_1} A \cos(h_1 y) \\
 \\
 E_z^0 = A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 H_x^0 = \frac{j\mathbf{w}\mathbf{e}_2}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 E_y^0 = -\frac{j\mathbf{b}}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 \\
 E_z^0 = -A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\
 H_x^0 = \frac{j\mathbf{w}\mathbf{e}_2}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\
 E_y^0 = -\frac{j\mathbf{b}}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)}
 \end{array}
 \right.$$

$$\mathbf{n} = \frac{\mathbf{e}_2}{\mathbf{e}_1} h_1 \tan\left(\frac{h_1 b}{2}\right)$$

$$\sqrt{\left(\frac{\mathbf{w}}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2} = \frac{\mathbf{e}_2}{\mathbf{e}_1} h_1 \tan\left(\frac{h_1 b}{2}\right)$$

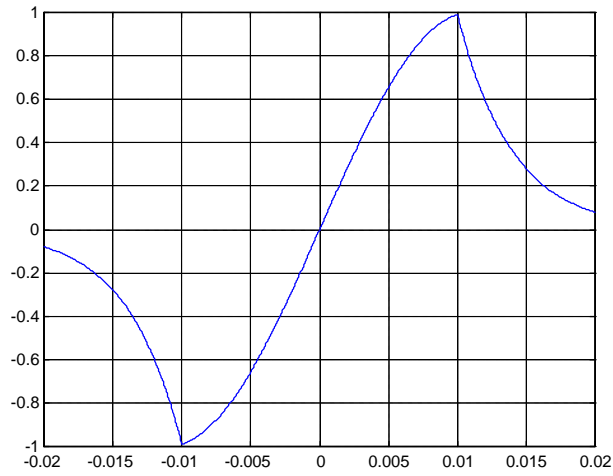
# $TM_{impar,1}$

---

$$f = 8\text{GHz}$$

$$h_{1,1} = 143.01$$

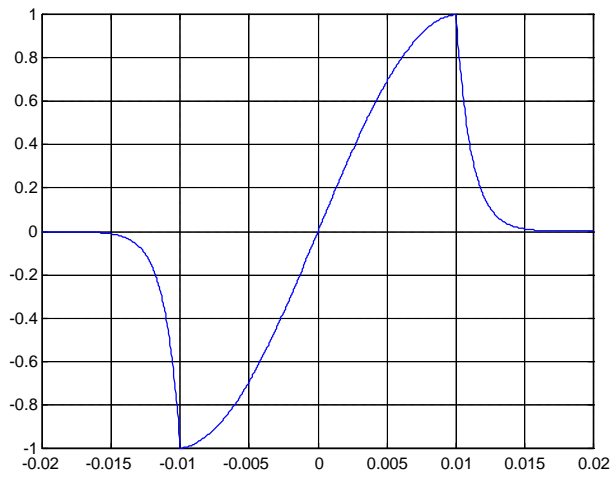
$$n_1 = 252.52$$



$$f = 25\text{GHz}$$

$$h_{1,1} = 152.81$$

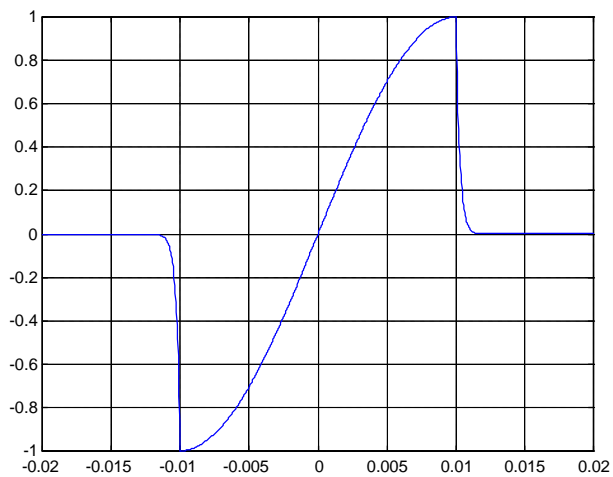
$$n_1 = 893.93$$



$$f = 100\text{GHz}$$

$$h_{1,1} = 156.0$$

$$n_1 = 3624.24$$



## Frequência de corte

$$n = 0 \quad \longrightarrow \quad \frac{e_2}{e_1} h_1 \tan\left(\frac{h_1 b}{2}\right) = 0$$

$$\frac{h_1 b}{2} = (n-1)\mathbf{p} \quad \Leftrightarrow \quad h_1 = \frac{(n-1)2\mathbf{p}}{b}, \quad n = 1, 2, \dots$$

$$(h_1)_{\text{corte}} = \frac{\mathbf{w}}{c} \sqrt{n_1^2 - n_2^2} \quad \longrightarrow$$

$$(f_c)_{TM \text{ ímpar}} = \frac{(n-1)c}{b\sqrt{n_1^2 - n_2^2}}, \quad n = 1, 2, \dots$$



## Ondas TE pares

$$E_z^0 = 0 \quad H_z^0 = \mathbf{y}_{par}$$

$$|y| \leq \frac{b}{2} : \left\{ \begin{array}{l} H_z^0 = B \cos(h_1 y) \\ H_y^0 = \frac{j\mathbf{b}}{h_1} B \sin(h_1 y) \\ E_x^0 = \frac{j\mathbf{w}\mathbf{m}_0}{h_1} B \sin(h_1 y) \end{array} \right.$$

$$y > \frac{b}{2} : \left\{ \begin{array}{l} H_z^0 = B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\ H_y^0 = -\frac{j\mathbf{b}}{\mathbf{n}} B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\ E_x^0 = -\frac{j\mathbf{w}\mathbf{m}_0}{\mathbf{n}} B \cos\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \end{array} \right.$$

$$y < -\frac{b}{2} : \left\{ \begin{array}{l} H_z^0 = B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\ H_y^0 = \frac{j\mathbf{b}}{\mathbf{n}} B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\ E_x^0 = \frac{j\mathbf{w}\mathbf{m}_0}{\mathbf{n}} B \cos\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \end{array} \right.$$

$$\sqrt{\left(\frac{\mathbf{w}}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2} = -h_1 \cot\left(\frac{h_1 b}{2}\right)$$

$$(f_c)_{TE\ par} = \frac{\left(n - \frac{1}{2}\right)c}{b\sqrt{n_1^2 - n_2^2}}, \quad n = 1, 2, \dots$$

$$(f_c)_{TE\ par} = (f_c)_{TM\ par}$$

## Modos TE ímpares

$$E_z^0 = 0$$

$$H_z^0 = \mathbf{y}_{\text{ímpar}}$$

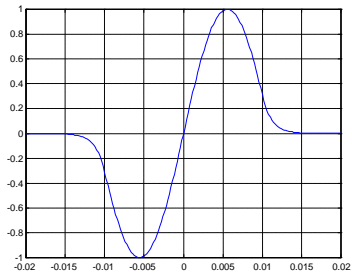
$$\begin{array}{l}
 |y| \leq \frac{b}{2}: \\
 \\
 y > \frac{b}{2}: \\
 \\
 y < -\frac{b}{2}:
 \end{array}
 \left\{
 \begin{array}{l}
 H_z^0 = A \sin(h_1 y) \\
 H_y^0 = -\frac{j\mathbf{b}}{h_1} A \cos(h_1 y) \\
 E_x^0 = -\frac{j\mathbf{w}\mathbf{m}_0}{h_1} A \cos(h_1 y) \\
 \\
 H_z^0 = A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 H_y^0 = -\frac{j\mathbf{b}}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 E_x^0 = -\frac{j\mathbf{w}\mathbf{m}_0}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{-n\left(y-\frac{b}{2}\right)} \\
 \\
 H_z^0 = -A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\
 H_y^0 = -\frac{j\mathbf{b}}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)} \\
 E_x^0 = -\frac{j\mathbf{w}\mathbf{m}_0}{n} A \sin\left(\frac{h_1 b}{2}\right) e^{n\left(y+\frac{b}{2}\right)}
 \end{array}
 \right.$$

$$\sqrt{\left(\frac{\mathbf{w}}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2} = -h_1 \tan\left(\frac{h_1 b}{2}\right)$$

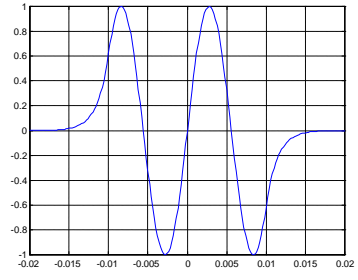
$$(f_c)_{TE \text{ ímpar}} = \frac{(n-1)c}{b\sqrt{n_1^2 - n_2^2}}, \quad n = 1, 2, \dots$$

$$(f_c)_{TE \text{ ímpar}} = (f_c)_{TM \text{ ímpar}}$$

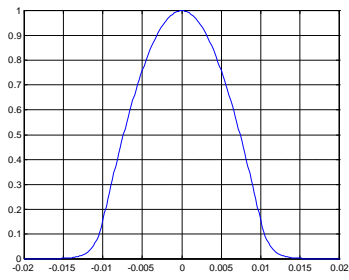
$TE_{par, 1}$



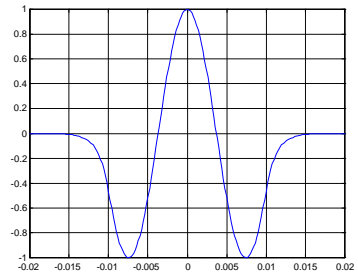
$TE_{par, 2}$



$TE_{impar, 1}$

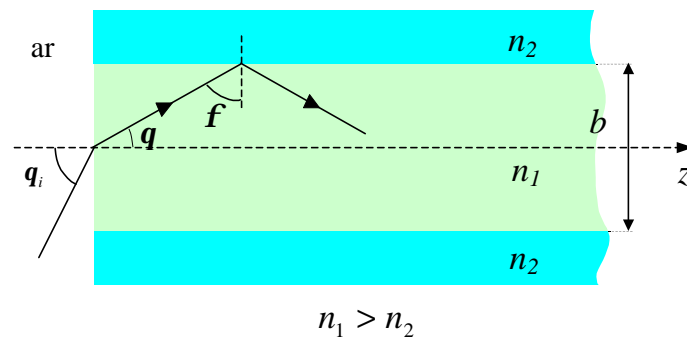


$TE_{impar, 2}$



MODOS		RELAÇÃO CARACTERÍSTICA	FREQUÊNCIA DE CORTE
PARES	TM	$\mathbf{n} = -\left(\frac{n_2}{n_1}\right)^2 h_1 \cot\left(\frac{h_1 b}{2}\right)$	$f_c = \frac{\left(n - \frac{1}{2}\right)c}{b\sqrt{n_1^2 - n_2^2}}$
	TE	$\mathbf{n} = -h_1 \cot\left(\frac{h_1 b}{2}\right)$	
ÍMPARES	TM	$\mathbf{n} = \left(\frac{n_2}{n_1}\right)^2 h_1 \tan\left(\frac{h_1 b}{2}\right)$	$f_c = \frac{(n-1)c}{b\sqrt{n_1^2 - n_2^2}}$
	TE	$\mathbf{n} = h_1 \tan\left(\frac{h_1 b}{2}\right)$	

## Guias dielétricos e reflexão interna total



$$n_1 \cos(\mathbf{f}) = \sin(\mathbf{q}_i)$$

**interface 1-2:**

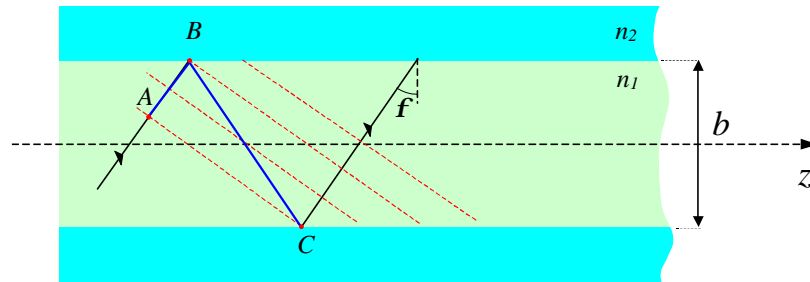
$$\mathbf{f} > \mathbf{f}_c = \sin^{-1}\left(\frac{n_2}{n_1}\right) \quad \Rightarrow \quad \text{Reflexão interna total}$$



$$\text{onda guiada} \quad \Rightarrow \quad \mathbf{f} \geq \mathbf{f}_c \quad \Rightarrow \quad \boxed{\sin(\mathbf{q}_i) \leq \sqrt{n_1^2 - n_2^2}}$$

- $\sqrt{n_1^2 - n_2^2}$        $\longrightarrow$     **abertura numérica (NA)**
- $\mathbf{q}_A = \sin^{-1}(NA)$      $\longrightarrow$     **ângulo de aceitação**

## Modos permitidos



- A e C  $\in$  mesma frente de onda  $\implies \mathbf{d}_C - \mathbf{d}_A = 2n\mathbf{p}$

- $\mathbf{d}_C = \mathbf{d}_A + k_1 l_{AB} + \mathbf{d}_r + k_1 l_{BC} + \mathbf{d}_r$  ,  $k_1 = \frac{\mathbf{w}n_1}{c}$



$$k_1(l_{AB} + l_{BC}) + 2\mathbf{d}_r = 2n\mathbf{p}$$

- $b = l_{BC} \cos(\mathbf{f}) \implies l_{BC} = \frac{b}{\cos(\mathbf{f})}$

- $l_{AB} = l_{BC} \sin(90^\circ - 2\mathbf{f}) = l_{BC} \cos(2\mathbf{f}) = \frac{b \cos(2\mathbf{f})}{\cos(\mathbf{f})}$

$$\implies 2k_1 b \cos(\mathbf{f}) + 2\mathbf{d}_r = 2n\mathbf{p} , n \text{ inteiro}$$

## Determinação de $d_r$

- polarização  $\perp$   $\iff$  ondas TE
- polarização  $\parallel$   $\iff$  ondas TM

$$\Gamma_{TE} = \frac{n_1 \cos(\mathbf{f}) - \sqrt{n_2^2 - n_1^2 \sin^2(\mathbf{f})}}{n_1 \cos(\mathbf{f}) + \sqrt{n_2^2 - n_1^2 \sin^2(\mathbf{f})}}$$

$$\Gamma_{TM} = \frac{-n_2^2 \cos(\mathbf{f}) + n_1 \sqrt{n_2^2 - n_1^2 \sin^2(\mathbf{f})}}{n_2^2 \cos(\mathbf{f}) + n_1 \sqrt{n_2^2 - n_1^2 \sin^2(\mathbf{f})}}$$

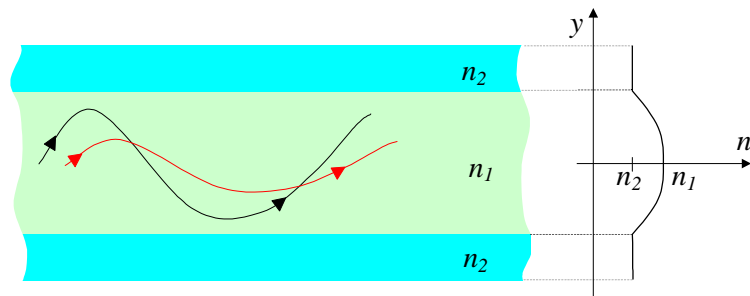
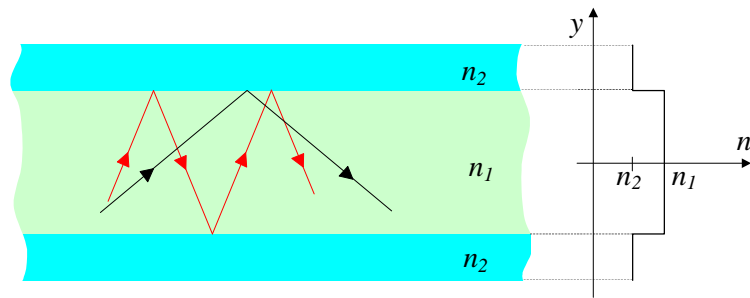
$$\mathbf{f} > \mathbf{f}_c$$
$$n_1^2 \sin^2(\mathbf{f}) > n_2^2$$

$$\Gamma = e^{jd_r}$$

$$d_{r,TE} = -2 \tan^{-1} \left( \frac{\sqrt{n_1^2 \sin^2(\mathbf{f}) - n_2^2}}{n_1 \cos(\mathbf{f})} \right)$$

$$d_{r,TM} = \mathbf{p} - 2 \tan^{-1} \left( \frac{n_1 \sqrt{n_1^2 \sin^2(\mathbf{f}) - n_2^2}}{n_2^2 \cos(\mathbf{f})} \right)$$

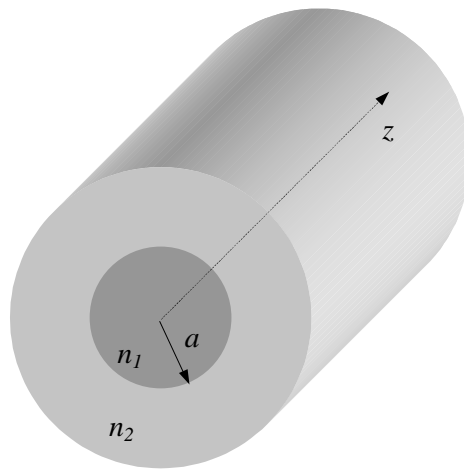
# Índice de refração gradual





# GUIAS DIELECTRICOS CIRCULARES

## GUIAS DIELÉTRICAS CIRCULARES



núcleo

$$\begin{cases} \nabla_{rf}^2 E_z^0 + h_1^2 E_z^0 = 0 \\ \nabla_{rf}^2 H_z^0 + h_1^2 H_z^0 = 0 \end{cases}$$

$$h_1^2 = \mathbf{g}^2 + \left( \frac{\mathbf{w}n_1}{c} \right)^2$$

bainha

$$\begin{cases} \nabla_{rf}^2 E_z^0 + h_2^2 E_z^0 = 0 \\ \nabla_{rf}^2 H_z^0 + h_2^2 H_z^0 = 0 \end{cases}$$

$$h_2^2 = \mathbf{g}^2 + \left( \frac{\mathbf{w}n_2}{c} \right)^2$$

## Equação de onda em guias dielétricos circulares

$$\nabla_{r\mathbf{f}}^2 \mathbf{y} + h^2 \mathbf{y} = 0 \quad h^2 = \begin{cases} h_1^2, & r \leq a \\ h_2^2, & r > a \end{cases}$$

$$\mathbf{y}(r, \mathbf{f}) = R(r)\Phi(\mathbf{f})$$
$$\Phi(\mathbf{f}) = A e^{j\mathbf{n}\mathbf{f}}$$

$$r^2 \frac{d^2 R(r)}{dr^2} + r \frac{dR(r)}{dr} + (h^2 r^2 - n^2) R(r) = 0$$

•  $h^2 > 0 \Leftrightarrow h$  real

$$R(r) = B J_n(hr)$$

•  $h^2 < 0 \Leftrightarrow h = j\mathbf{n}$

$$r^2 \frac{d^2 R(r)}{dr^2} + r \frac{dR(r)}{dr} - (\mathbf{n}^2 r^2 + n^2) R(r) = 0$$

equação diferencial  
de Bessel modificada

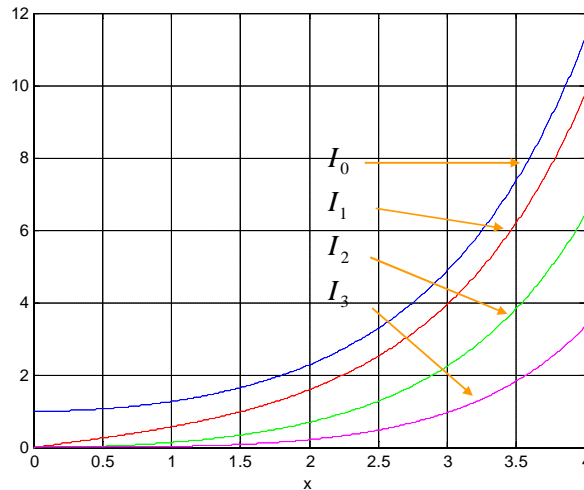
$$R(r) = C I_n(\mathbf{n} r) + D K_n(\mathbf{n} r)$$

funções de Bessel  
modificadas de 1ª e  
2ª espécies

## Funções de Bessel modificadas de 1ª espécie

Para  $n$  inteiro

$$I_n(x) = j^{-n} J_n(jx) = \sum_{k=0}^{\infty} \frac{(x/2)^{n+2k}}{k!(n+k)!}$$



- $\lim_{x \rightarrow \infty} I_n(x) = \infty$

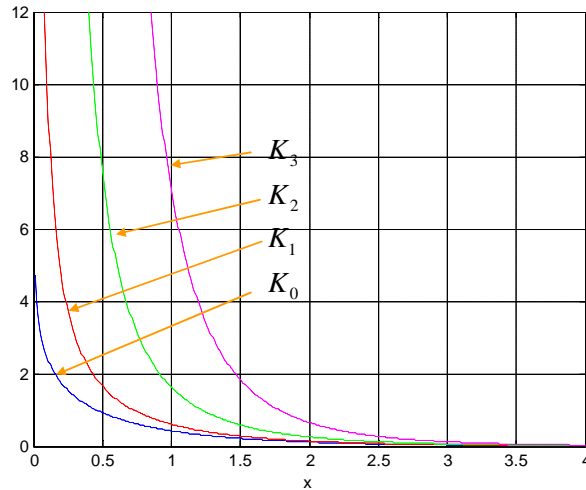


$I_n(x)$  não deve fazer parte da solução geral quando a região de interesse incluir o infinito!

## Funções de Bessel modificadas de 2ª espécie

Para  $n$  inteiro

$$K_n(x) = \lim_{p \rightarrow n} \frac{p}{2 \sin(pp)} [I_{-p}(x) - I_p(x)]$$



- $\lim_{x \rightarrow 0} K_n(x) = \infty$



$K_n(x)$  não deve fazer parte da solução geral quando a região de interesse incluir a origem!

onda guiada  $\Rightarrow h_1$  real e  $h_2 = j\mathbf{n}$



$$\mathbf{n} = \sqrt{\left(\frac{\mathbf{w}}{c}\right)^2 (n_1^2 - n_2^2) - h_1^2}$$

•  $\mathbf{b} = \sqrt{\left(\frac{\mathbf{w}}{c} n_1\right)^2 - h_1^2} = \sqrt{\left(\frac{\mathbf{w}}{c} n_2\right)^2 + \mathbf{n}^2} \Rightarrow \frac{\mathbf{w}}{c} n_1 < \mathbf{b} < \frac{\mathbf{w}}{c} n_2$

$\Rightarrow$  
$$\mathbf{y}(r, \mathbf{f}) = \begin{cases} AJ_n(h_1 r) e^{j\mathbf{n}f}, & r \leq a \\ BK_n(\mathbf{n} r) e^{j\mathbf{n}f}, & r > a \end{cases}$$

• núcleo: 
$$E_z^0 = AJ_n(h_1 r) e^{j\mathbf{n}f}$$
$$H_z^0 = BJ_n(h_1 r) e^{j\mathbf{n}f}$$

• bainha: 
$$E_z^0 = CK_n(\mathbf{n} r) e^{j\mathbf{n}f}$$
$$H_z^0 = DK_n(\mathbf{n} r) e^{j\mathbf{n}f}$$

- núcleo

$$\begin{aligned}
 E_z^0 &= AJ_n(h_1 r) e^{jn f} \\
 H_z^0 &= BJ_n(h_1 r) e^{jn f} \\
 H_r^0 &= -\frac{1}{h_1^2} \left[ j b h_1 B J'_n(h_1 r) + \frac{w e_1 n}{r} A J_n(h_1 r) \right] e^{jn f} \\
 H_f^0 &= -\frac{1}{h_1^2} \left[ -\frac{b n}{r} B J_n(h_1 r) + j w e_1 h_1 A J'_n(h_1 r) \right] e^{jn f} \\
 E_r^0 &= -\frac{1}{h_1^2} \left[ j b h_1 A J'_n(h_1 r) - \frac{w m_0 n}{r} B J_n(h_1 r) \right] e^{jn f} \\
 E_f^0 &= -\frac{1}{h_1^2} \left[ -\frac{b n}{r} A J_n(h_1 r) - j w m_0 h_1 B J'_n(h_1 r) \right] e^{jn f}
 \end{aligned}$$

- bainha

$$\begin{aligned}
 E_z^0 &= CK_n(n r) e^{jn f} \\
 H_z^0 &= DK_n(n r) e^{jn f} \\
 H_r^0 &= \frac{1}{n^2} \left[ j b n D K'_n(n r) + \frac{w e_2 n}{r} C K_n(n r) \right] e^{jn f} \\
 H_f^0 &= \frac{1}{n^2} \left[ -\frac{b n}{r} D K_n(n r) + j w e_2 n C K'_n(n r) \right] e^{jn f} \\
 E_r^0 &= \frac{1}{n^2} \left[ j b n C K'_n(n r) - \frac{w m_0 n}{r} D K_n(n r) \right] e^{jn f} \\
 E_f^0 &= \frac{1}{n^2} \left[ -\frac{b n}{r} C K_n(n r) - j w m_0 n D K'_n(n r) \right] e^{jn f}
 \end{aligned}$$

### Condições fronteira

$$\begin{aligned}
 E_z^0 \text{ e } E_f^0 &\text{ contínuos em } r = a \\
 H_z^0 \text{ e } H_f^0 &\text{ contínuos em } r = a
 \end{aligned}$$



$$\begin{aligned}
 AJ_n(h_1 a) &= CK_n(\mathbf{n} a) \Leftrightarrow AJ_n(h_1 a) - CK_n(\mathbf{n} a) = 0 \\
 BJ_n(h_1 a) - DK_n(\mathbf{n} a) &= 0 \\
 B \frac{\mathbf{b} n}{h_1^2 a} J_n(h_1 a) - A \frac{j\mathbf{w} \mathbf{e}_1}{h_1} J'_n(h_1 a) + D \frac{\mathbf{b} n}{\mathbf{n}^2 a} K_n(\mathbf{n} a) - C \frac{j\mathbf{w} \mathbf{e}_2}{\mathbf{n}} K'_n(\mathbf{n} a) &= 0 \\
 A \frac{\mathbf{b} n}{h_1^2 a} J_n(h_1 a) + B \frac{j\mathbf{w} \mathbf{m}_0}{h_1} J'_n(h_1 a) + C \frac{\mathbf{b} n}{\mathbf{n}^2 a} K_n(\mathbf{n} a) + D \frac{j\mathbf{w} \mathbf{m}_0}{\mathbf{n}} K'_n(\mathbf{n} a) &= 0
 \end{aligned}$$

solução não trivial



$$\begin{vmatrix}
 J_n(h_1 a) & 0 & K_n(\mathbf{n} a) & 0 \\
 0 & J_n(h_1 a) & 0 & K_n(\mathbf{n} a) \\
 -\frac{j\mathbf{w} \mathbf{e}_1}{h_1} J'_n(h_1 a) & \frac{\mathbf{b} n}{h_1^2 a} J_n(h_1 a) & -\frac{j\mathbf{w} \mathbf{e}_2}{\mathbf{n}} K'_n(\mathbf{n} a) & \frac{\mathbf{b} n}{\mathbf{n}^2 a} K_n(\mathbf{n} a) \\
 \frac{\mathbf{b} n}{h_1^2 a} J_n(h_1 a) & \frac{j\mathbf{w} \mathbf{m}_0}{h_1} J'_n(h_1 a) & \frac{\mathbf{b} n}{\mathbf{n}^2 a} K_n(\mathbf{n} a) & \frac{j\mathbf{w} \mathbf{m}_0}{\mathbf{n}} K'_n(\mathbf{n} a)
 \end{vmatrix} = 0$$



$$\left( \frac{\mathbf{w}}{c} \right)^2 \left[ \frac{J'_n(h_1 a)}{h_1 J_n(h_1 a)} + \frac{K'_n(\mathbf{n} a)}{\mathbf{n} K_n(\mathbf{n} a)} \right] \left[ n_1^2 \frac{J'_n(h_1 a)}{h_1 J_n(h_1 a)} + n_2^2 \frac{K'_n(\mathbf{n} a)}{\mathbf{n} K_n(\mathbf{n} a)} \right] = \left( \frac{\mathbf{b} n}{a} \right)^2 \left( \frac{1}{h_1^2} + \frac{1}{\mathbf{n}^2} \right)^2$$



**equação característica para os modos TM, TE, EH e HE**



$n=0$

$$\begin{bmatrix} J_0(h_1 a) & 0 & K_0(\mathbf{n} a) & 0 \\ 0 & J_0(h_1 a) & 0 & K_0(\mathbf{n} a) \\ -\frac{j\omega \mathbf{e}_1}{h_1} J'_0(h_1 a) & 0 & -\frac{j\omega \mathbf{e}_2}{\mathbf{n}} K'_0(\mathbf{n} a) & 0 \\ 0 & \frac{j\omega \mathbf{m}_0}{h_1} J'_0(h_1 a) & 0 & \frac{j\omega \mathbf{m}_0}{\mathbf{n}} K'_0(\mathbf{n} a) \end{bmatrix} \begin{bmatrix} A \\ B \\ C \\ D \end{bmatrix} = 0$$

$$\begin{bmatrix} J_0(h_1 a) & K_0(\mathbf{n} a) \\ -\frac{\mathbf{e}_1}{h_1} J'_0(h_1 a) & -\frac{\mathbf{e}_2}{\mathbf{n}} K'_0(\mathbf{n} a) \end{bmatrix} \begin{bmatrix} A \\ C \end{bmatrix} = 0$$

e

$$\begin{bmatrix} J_0(h_1 a) & K_0(\mathbf{n} a) \\ \frac{1}{h_1} J'_0(h_1 a) & \frac{1}{\mathbf{n}} K'_0(\mathbf{n} a) \end{bmatrix} \begin{bmatrix} B \\ D \end{bmatrix} = 0$$

solução não trivial  
 $B = D = 0$  possível  
**modos TM**

solução não trivial  
 $A = C = 0$  possível  
**modos TE**

$$\left[ n_1^2 \frac{J_1(h_1 a)}{h_1 J_0(h_1 a)} + n_2^2 \frac{K_1(\mathbf{n} a)}{\mathbf{n} K_0(\mathbf{n} a)} \right] = 0$$

$$\left[ \frac{J_1(h_1 a)}{h_1 J_0(h_1 a)} + \frac{K_1(\mathbf{n} a)}{\mathbf{n} K_0(\mathbf{n} a)} \right] = 0$$

## Frequência de corte

$n$	modo	condição de corte
0	TE <sub>0p</sub> TM <sub>0p</sub>	$J_0(h_1 a) = 0$
1	HE <sub>1p</sub> EH <sub>1p</sub>	$J_1(h_1 a) = 0$
$\geq 2$	EH <sub>np</sub>	$J_n(h_1 a) = 0$
	HE <sub>np</sub>	$\left(\frac{n_1^2}{n_2^2} + 1\right) J_{n-1}(h_1 a) = \frac{h_1 a}{n-1} J_n(h_1 a)$

- modo HE<sub>11</sub> tem frequência de corte nula
- modos seguintes: TE<sub>01</sub> e TM<sub>01</sub> (frequência de corte associada ao primeiro zero de  $J_0$ , em 2.405)
- parâmetro  $V$  ou frequência normalizada:

$$V^2 = (h_1^2 + n^2) a^2 = \left(\frac{w a}{c}\right)^2 (n_1^2 - n_2^2)$$



$$V = \frac{2p a}{I_0} \sqrt{n_1^2 - n_2^2}$$

- $V > 2.405$  multimodo
- $V \leq 2.405$  monomodo.